

ECE 2695: Adaptive Control (3 Credits, Fall 2008)

Lecture 3: System Identification (I)

September 15, 2008

Instructor: Zhi-Hong Mao
Assistant Professor of ECE and Bioengineering
University of Pittsburgh, Pittsburgh, PA

1

Outline

- Homework 2
- Review of last lecture
- Basic identification methods

2

Homework 1

- Problems 1.10, 1.12 (a)-(c), and 2.2 (a)-(b)
- Due on 9/22 (Monday) in class but before the lecture starts

3

Review of last lecture

- State-space description of dynamic systems

$$\frac{dx}{dt} = f(t, x), \quad x(t_0) = x_0$$

where $x \in \mathbf{R}^n$, $t \geq 0$.

- Autonomous and non-autonomous systems
- Equilibrium

Question: What is the equilibrium of the following system?

$$\dot{x} = Ax$$

4

Review of last lecture

- State-space description of dynamic systems

- Autonomous and non-autonomous systems
- Equilibrium

- Lipschitz condition and consequences

- Definition: The function f is said to be Lipschitz in x if, for some $h > 0$, there exists $l \geq 0$ such that

$$|f(t, x_1) - f(t, x_2)| \leq l|x_1 - x_2|$$

for all $x_1, x_2 \in B_h$ (closed ball of radius h centered at a point of interest in \mathbf{R}^n), $t > 0$. The constant l is called the Lipschitz constant. Globally Lipschitz functions satisfy the above inequality for all $x_1, x_2 \in \mathbf{R}^n$

5

Review of last lecture

- State-space description of dynamic systems

- Autonomous and non-autonomous systems
- Equilibrium

- Lipschitz condition and consequences

$$|f(t, x_1) - f(t, x_2)| \leq l|x_1 - x_2|$$

Remarks:

- (1) The Lipschitz property is by default assumed to be satisfied uniformly, i.e., l does not depend on t .
- (2) If f is Lipschitz in x , then it is continuous in x .
- (3) If f has continuous and bounded partial derivatives in x , then it is Lipschitz.

6

Review of last lecture

- State-space description of dynamic systems

- Autonomous and non-autonomous systems
- Equilibrium

- Lipschitz condition and consequences

- Definition

$$\frac{dx}{dt} = f(t, x), \quad x(t_0) = x_0 \quad (*)$$

- A sufficient condition for the **existence** and **uniqueness** of the solutions of (*) on some time interval (for as long as $x \in B_t$): f locally bounded and locally Lipschitz in x

7

Review of last lecture

- State-space description of dynamic systems

- Autonomous and non-autonomous systems
- Equilibrium

- Lipschitz condition and consequences

- Definition

- A sufficient condition for the existence and uniqueness of the solutions

- Examples

a) $f(t, x) = \frac{x}{1+x^2}$

c) $f(t, x) = Ax$

e) $f(t, x) = \frac{e^x - e^{-x}}{e^x + e^{-x}}$

What if x is a vector?

b) $f(t, x) = \frac{e^t}{1+e^t} x$

d) $f(t, x) = e^t x$

f) $f(t, x) = \sqrt{x}$

8

Review of last lecture

- State-space description of dynamic systems

- Stability of dynamic systems

- Stability in LTI systems

The zero-input response of the following system

$$\dot{x} = Ax + Bu$$

$$y = Cx + Du$$

or the equation

$$\dot{x} = Ax$$

is **marginally stable** or **stable in the sense of Lyapunov** if every finite initial state excites a bounded response. It is **asymptotically stable** if every finite initial state excites a bounded response, which in addition, approaches **0** as t approaches ∞

9

Review of last lecture

- State-space description of dynamic systems

- **Stability of dynamic systems**

- Stability in LTI systems

If A has distinct eigenvalues, then the equation $\dot{x} = Ax$ is marginally stable if and only if all eigenvalues of A have zero or negative real parts.

The equation $\dot{x} = Ax$ is marginally stable if and only if all eigenvalues of A have zero or negative real parts and those with zero real parts are simple roots of the minimal polynomial of A .

The equation $\dot{x} = Ax$ is asymptotically stable if and only all eigenvalues of A have negative real parts.

10

Review of last lecture

- State-space description of dynamic systems

- **Stability of dynamic systems**

- Stability in LTI systems

- Examples

$$A_1 = \begin{bmatrix} -2 & 0 \\ 0 & -3 \end{bmatrix} \quad A_2 = \begin{bmatrix} 0 & 0 \\ 0 & -1 \end{bmatrix}$$
$$A_3 = \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} \quad A_4 = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix} \quad A_5 = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}$$

11

Review of last lecture

- State-space description of dynamic systems

- **Stability of dynamic systems**

- Stability in LTI systems

$$\frac{dx}{dt} = f(t, x), \quad x(t_0) = x_0 \quad (*)$$

- Stability in the sense of Lyapunov

- The point $x = 0$ is called a **stable** equilibrium point of (*), if, for all $t_0 \geq 0$ and $\epsilon > 0$, there exists $\delta(t_0, \epsilon)$ such that

$$|x_0| < \delta(t_0, \epsilon) \Rightarrow |x(t)| < \epsilon \quad \text{for all } t \geq t_0$$

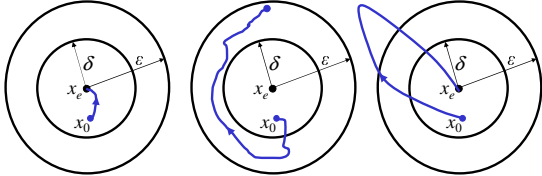
where $x(t)$ is the solution of (*) starting from x_0 at t_0

Question: How to generalize the above definition to a nonzero equilibrium?

12

Review of last lecture

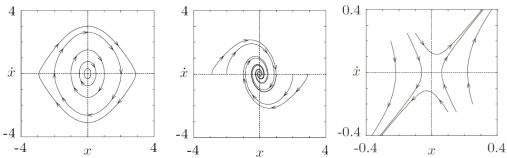
- State-space description of dynamic systems
- **Stability of dynamic systems**
 - Stability in LTI systems
 - Stability in the sense of Lyapunov



13

Review of last lecture

- State-space description of dynamic systems
- **Stability of dynamic systems**
 - Stability in LTI systems
 - Stability in the sense of Lyapunov
 - Examples



14

Review of last lecture

- State-space description of dynamic systems
- **Stability of dynamic systems**
 - Stability in LTI systems
 - Stability in the sense of Lyapunov
 - Uniform stability
 - Asymptotic stability
 - Uniform asymptotic stability
 - Global asymptotic stability
 - Exponential stability
 - Global exponential stability

15

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems
- Lyapunov stability theory
 - Basic idea: If some “measure of the energy” associated with a system is decreasing, then the system will converge to its equilibrium

16

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems
- Lyapunov stability theory
 - Basic idea
 - A few concepts
 - Class K functions: A function $\alpha(\cdot): \mathbf{R}_+ \rightarrow \mathbf{R}_+$ belongs to class K (denoted $\alpha(\cdot) \in K$), if it is continuous, strictly increasing, and $\alpha(0) = 0$

17

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems
- Lyapunov stability theory
 - Basic idea
 - A few concepts
 - Class K functions
 - Locally positive definite functions (l.p.d.f.): A continuous function $v(t, x): \mathbf{R}_+ \times \mathbf{R}^n \rightarrow \mathbf{R}_+$ is called a locally positive definite function (l.p.d.f.) if, for some $h > 0$, and some $\alpha(\cdot) \in K$
$$v(t, 0) = 0 \quad \text{and} \quad v(t, x) \geq \alpha(|x|) \quad \text{for all } x \in B_h, t \geq 0$$

18

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

- **Lyapunov stability theory**

- Basic idea

- **A few concepts**

- Class K functions
 - Locally positive definite functions (l.p.d.f.)

- **Positive definite functions (p.d.f.):** A continuous function $v(t, x): \mathbf{R}_+ \times \mathbf{R}^n \rightarrow \mathbf{R}_+$ is called a positive definite function (p.d.f.) if, for some $\alpha(\cdot) \in K$

$$v(t, 0) = 0 \quad \text{and} \quad v(t, x) \geq \alpha(|x|) \quad \text{for all } x \in \mathbf{R}^n, t \geq 0$$

19

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

- **Lyapunov stability theory**

- Basic idea

- **A few concepts**

- Class K functions
 - Locally positive definite functions (l.p.d.f.)

- **Positive definite functions (p.d.f.)**

Question: Are these functions p.d.f.?

$$v(t, x) = |x|^2 \quad v(t, x) = x^T P x, \text{ with } P > 0$$

$$v(t, x) = (t + 1)|x|^2 \quad v(t, x) = e^{-t}|x|^2$$

$$v(t, x) = \sin^2(|x|^2)$$

20

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

- **Lyapunov stability theory**

- Basic idea

- **A few concepts**

- Class K functions
 - Locally positive definite functions (l.p.d.f.)
 - Positive definite functions (p.d.f.)

- **Decrescent functions:** The function $v(t, x)$ is called decrescent, if there exists a function $\beta(\cdot) \in K$, such that

$$v(t, x) \leq \beta(|x|) \quad \text{for all } x \in B_h, t \geq 0$$

21

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

• Lyapunov stability theory

– Basic idea

– A few concepts

- Class K functions
- Locally positive definite functions (l.p.d.f.)
- Positive definite functions (p.d.f.)
- Decrescent functions

Question: Are these functions decrescent functions?

$$v(t, x) = |x|^2 \quad v(t, x) = x^T P x, \text{ with } P > 0$$

$$v(t, x) = (t + 1)|x|^2 \quad v(t, x) = e^{-t}|x|^2$$

$$v(t, x) = \sin^2(|x|^2)$$

22

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

• Lyapunov stability theory

– Basic idea

– A few concepts

– Basic theorem of Lyapunov

- Let $v(t, x)$ be continuously differentiable. Then

Conditions on $v(t, x)$	Conditions on $-dv(t, x)/dt$	Conclusions
l.p.d.f.	≥ 0 locally	stable
l.p.d.f., decrescent	≥ 0 locally	uniformly stable
l.p.d.f.	l.p.d.f.	asymptotically stable
l.p.d.f., decrescent	l.p.d.f.	uniformly asymptotically stable
p.d.f., decrescent	p.d.f.	globally u.a.s.

23

$$\frac{dx}{dt} = f(t, x), \quad x(t_0) = x_0 \quad (*)$$

$$\frac{dv(t, x)}{dt} = \frac{\partial v(t, x)}{\partial t} + \frac{\partial v(t, x)}{\partial x} f(t, x)$$

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

• Lyapunov stability theory

– Basic idea

– A few concepts

– Basic theorem of Lyapunov

- Examples

$$\dot{x}_1 = x_1(x_1^2 + x_2^2 - 1) - x_2$$

$$\dot{x}_2 = x_1 + x_2(x_1^2 + x_2^2 - 1)$$

$$v(x_1, x_2) = x_1^2 + x_2^2$$

24

Review of last lecture

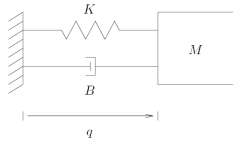
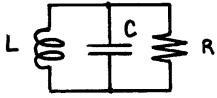
- State-space description of dynamic systems
- Stability of dynamic systems

- **Lyapunov stability theory**

- Basic idea
- A few concepts

- **Basic theorem of Lyapunov**

- Examples



25

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

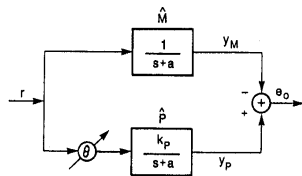
- **Lyapunov stability theory**

- Basic idea
- A few concepts

- **Basic theorem of Lyapunov**

- Examples

MRAC using the MIT rule and the Lyapunov redesign



26

Review of last lecture

- State-space description of dynamic systems
- Stability of dynamic systems

- **Lyapunov stability theory**

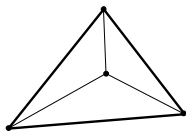
- Basic idea
- A few concepts

- **Basic theorem of Lyapunov**

- Examples

Question (Steiner's problem): How to find a point inside a triangle that gives the shortest sum of distances to the vertices?

Energy-function based optimization algorithms (with applications in protein folding problems, Hopfield neural networks, robotic path planning, etc.)



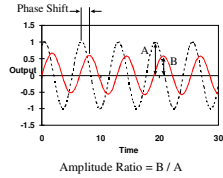
27

Basic identification methods

- Frequency domain approach
 - Frequency response: steady-state response of systems to sinusoidal inputs

The figure compares the output response of a system (red solid line) with a sinusoidal input (black dashed line)

Both the **magnitude** and the **phase shift** of a system will change with the frequency of the input into the system



28

Basic identification methods

- Frequency domain approach
 - Frequency response: steady-state response of systems to sinusoidal inputs
 - Frequency response function
 - Given a system with transfer function $\hat{P}(s)$, its frequency response function is $\hat{P}(j\omega)$
 - The steady-state gain of a system for a sinusoidal input $\sin(\omega_p t)$ is the **magnitude** of the transfer function evaluation at $s = j\omega_p$, and the **phase shift** of the output sinusoid relative to the input sinusoid is the angle of $\hat{P}(j\omega_p)$

29

Basic identification methods

- Frequency domain approach
 - Frequency response: steady-state response of systems to sinusoidal inputs
 - Frequency response function
 - Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p}$$

30

Basic identification methods

- Frequency domain approach

- Frequency response: steady-state response of systems to sinusoidal inputs
- Frequency response function
- Example

Question: How about identification of a system of higher order?

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{\alpha_n s^{n-1} + \dots + \alpha_1}{s^n + \beta_n s^{n-1} + \dots + \beta_1}$$

31

Basic identification methods

- Frequency domain approach

- Time domain approach

- Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p}$$

Question: What is the differential equation for the above frequency-domain description?

32

Basic identification methods

- Frequency domain approach

- Time domain approach

- Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p}$$

$$\dot{y}_p(t) = -a_p y_p(t) + k_p r(t)$$

Question: If we have measurement of $r(t)$ and $dy_p(t)/dt$ at t_1 and t_2 , how can we estimate a_p and k_p ?

33

Basic identification methods

• Frequency domain approach

• Time domain approach

– Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p}$$



$$\dot{y}_p(t) = -a_p y_p(t) + k_p r(t)$$

$$\begin{bmatrix} -a_p \\ k_p \end{bmatrix} = \begin{bmatrix} y_p(t_1) & r(t_1) \\ y_p(t_2) & r(t_2) \end{bmatrix}^{-1} \begin{bmatrix} \dot{y}_p(t_1) \\ \dot{y}_p(t_2) \end{bmatrix}$$

34

Basic identification methods

• Frequency domain approach

• Time domain approach

– Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p} \rightarrow \dot{y}_p(t) = -a_p y_p(t) + k_p r(t)$$

$$\begin{bmatrix} -a_p \\ k_p \end{bmatrix} = \begin{bmatrix} y_p(t_1) & r(t_1) \\ y_p(t_2) & r(t_2) \end{bmatrix}^{-1} \begin{bmatrix} \dot{y}_p(t_1) \\ \dot{y}_p(t_2) \end{bmatrix}$$

Question: Why we avoid the measurement of $dy_p(t)/dt$ and how to do that?

35

Basic identification methods

• Frequency domain approach

• Time domain approach

– Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p} \rightarrow \frac{s + a_p}{s + \lambda} \hat{y}_p = \frac{k_p}{s + \lambda} \hat{r}$$

Define $\hat{w}^{(1)} = \frac{1}{s + \lambda} \hat{r}$ $\hat{w}^{(2)} = \frac{1}{s + \lambda} \hat{y}_p$

$$\begin{aligned} y_p(t) &= k_p w^{(1)}(t) + (\lambda - a_p) w^{(2)}(t) \\ \dot{w}^{(1)} &= -\lambda w^{(1)} + r \\ \dot{w}^{(2)} &= -\lambda w^{(2)} + y_p \end{aligned}$$

36

Basic identification methods

- Frequency domain approach
- Time domain approach
 - Example

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p}$$

→

$$y_p(t) = k_p w^{(1)}(t) + (\lambda - a_p) w^{(2)}(t)$$

$$\dot{w}^{(1)} = -\lambda w^{(1)} + r$$

$$\dot{w}^{(2)} = -\lambda w^{(2)} + y_p$$

$$y_p(t) = \theta^{*T} w(t) = w^T(t) \theta^*$$

Define $\theta^* :=$

$$\begin{bmatrix} k_p \\ \lambda - a_p \end{bmatrix}$$

$$w(t) :=$$

$$\begin{bmatrix} w^{(1)}(t) \\ w^{(2)}(t) \end{bmatrix}$$

37

Basic identification methods

- Frequency domain approach
- Time domain approach
 - Example

$$y_p(t) = \theta^{*T} w(t) = w^T(t) \theta^*$$

↓

$$e_1(t) = \theta^T w(t) - y_p(t) = [\theta^T - \theta^{*T}] w(t)$$

θ^*
Nominal identifier parameter

θ
Adaptive identifier parameter

$e_1(t)$
Identification error

$[\theta^T - \theta^{*T}] w(t)$
Linear error equation

38

Basic identification methods

- Frequency domain approach
- Time domain approach
 - Example
 - Gradient algorithm

$$e_1(t) = \theta^T w(t) - y_p(t) = [\theta^T - \theta^{*T}] w(t)$$

Objective: minimize $e_1^2(t)$

Gradient: $\frac{\partial}{\partial \theta} (e_1^2) = 2e_1 \frac{\partial}{\partial \theta} (e_1) = 2e_1 w$

It is a vector!

Parameter update law: $\frac{d\theta}{dt} = -g e_1 w$

39

Basic identification methods

- Frequency domain approach
- Time domain approach
 - Example
 - Gradient algorithm
 - Advantages of using the gradient algorithm over using the following calculation

$$\begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix} = \begin{bmatrix} w_1(t_1) & w_2(t_1) \\ w_1(t_2) & w_2(t_2) \end{bmatrix}^{-1} \begin{bmatrix} y_p(t_1) \\ y_p(t_2) \end{bmatrix}$$

40

Basic identification methods

- Frequency domain approach
- Time domain approach
 - Example
 - Gradient algorithm
 - Least-squares algorithm

Objective: minimize the integral-squared-error (ISE)

$$\text{ISE} = \int_0^t [\theta^T(\tau)w(\tau) - y_p(\tau)]^2 d\tau$$

$$\hat{\theta}_{LS}(t) = \left[\int_0^t w(\tau)w^T(\tau) d\tau \right]^{-1} \left[\int_0^t w(\tau)y_p(\tau) d\tau \right]$$

Least-squares estimate

41

Basic identification methods

- Frequency domain approach
- Time domain approach
 - Example
 - Gradient algorithm
 - Least-squares algorithm

Recursive formulation

$$\begin{aligned} \dot{\theta}(t) &= -P(t)w(t)[\theta^T(t)w(t) - y_p(t)] & \theta(0) &= \theta_0 \\ \dot{P}(t) &= -P(t)w(t)w^T(t)P(t) & P(0) &= P^T(0) = P_0 \end{aligned}$$

Remark: It can be shown that $\hat{\theta}(t)$ converges asymptotically to θ^* if $\int_0^t w(\tau)w^T(\tau)d\tau$ is unbounded as $t \rightarrow \infty$.

42

Basic identification methods

- Frequency domain approach

- Time domain approach

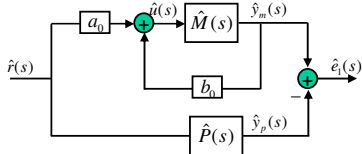
- Example
- Gradient algorithm
- Least-squares algorithm

- Model reference identification

Reference model

$$\frac{\hat{y}_p(s)}{\hat{r}(s)} = \hat{P}(s) = \frac{k_p}{s + a_p}$$

$$\frac{\hat{y}_m(s)}{\hat{u}(s)} = \hat{M}(s) = \frac{k_m}{s + a_m}$$



43

References

- K. J. Astrom and B. Wittenmark, Adaptive Control, 2nd Edition, Addison-Wesley, 1995.
- C.-T. Chen, Linear System Theory and Design, 3rd Edition, Oxford University Press, 1999.
- R. M. Murray, Z. Li, and S. S. Sastry, A Mathematical Introduction to Robotic Manipulation, CRC, 1994.
- S. Sastry and M. Bodson, Adaptive Control: Stability, Convergence, and Robustness, Prentice-Hall, 1989.

44
